

**VTR VIDEO SIGNAL PROCESSOR**

**DESCRIPTION**

The M51450P is a semiconductor integrated circuit designed for use in VTR video signal processing. It consists of a double limiter, FM demodulator, noise filter, Y/C mixer, sync clamp circuit, white clipper, squelch, playback video amplifier, video AGC, EE amplifier, recording video amplifier, E-E/V-V switch, and sync separation circuit.

**FEATURES**

- Packaged in a 30-pin DIL shrink package to minimize required PC board space
- Built-in double limiter
- To prevent misoperation of the TV set sync separation circuit, FM demodulator outputs below the sync level are shifted to white and clipped using a built-in circuit.
- Peak-type video AGC circuit
- Used in conjunction with the M51451P, all major VTR video signal processing functions may be performed.

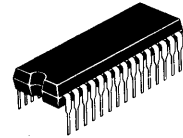
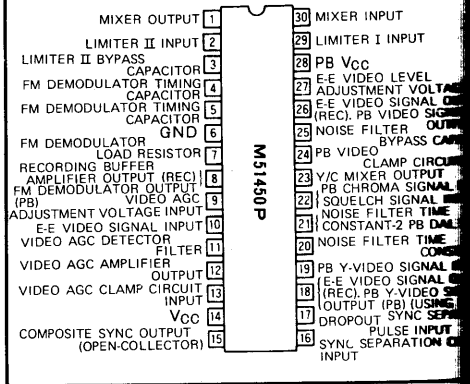
**APPLICATIONS**

VTR Video signal processing

**RECOMMENDED OPERATING CONDITIONS**

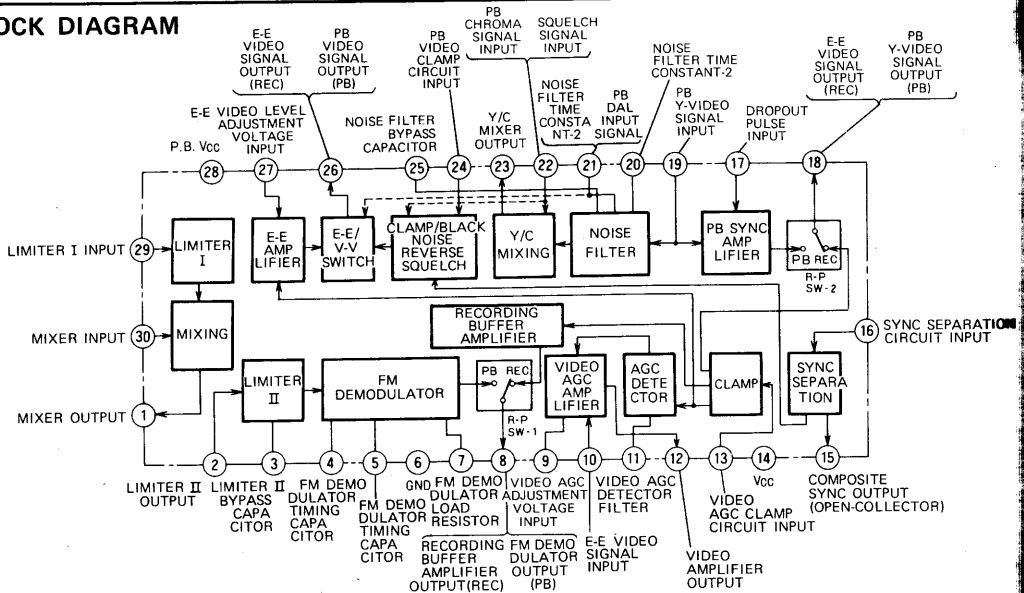
Supply voltage range ..... 8~10V  
 Rated supply voltage ..... 9V

**PIN CONFIGURATION (TOP VIEW)**



30-pin molded plastic DIL (shrink)

**BLOCK DIAGRAM**



MITSUBISHI LINEAR ICs  
**M51450P**

**VTR VIDEO SIGNAL PROCESSOR**

**ABSOLUTE MAXIMUM RATINGS** ( $T_a = 25^\circ\text{C}$ , unless otherwise noted)

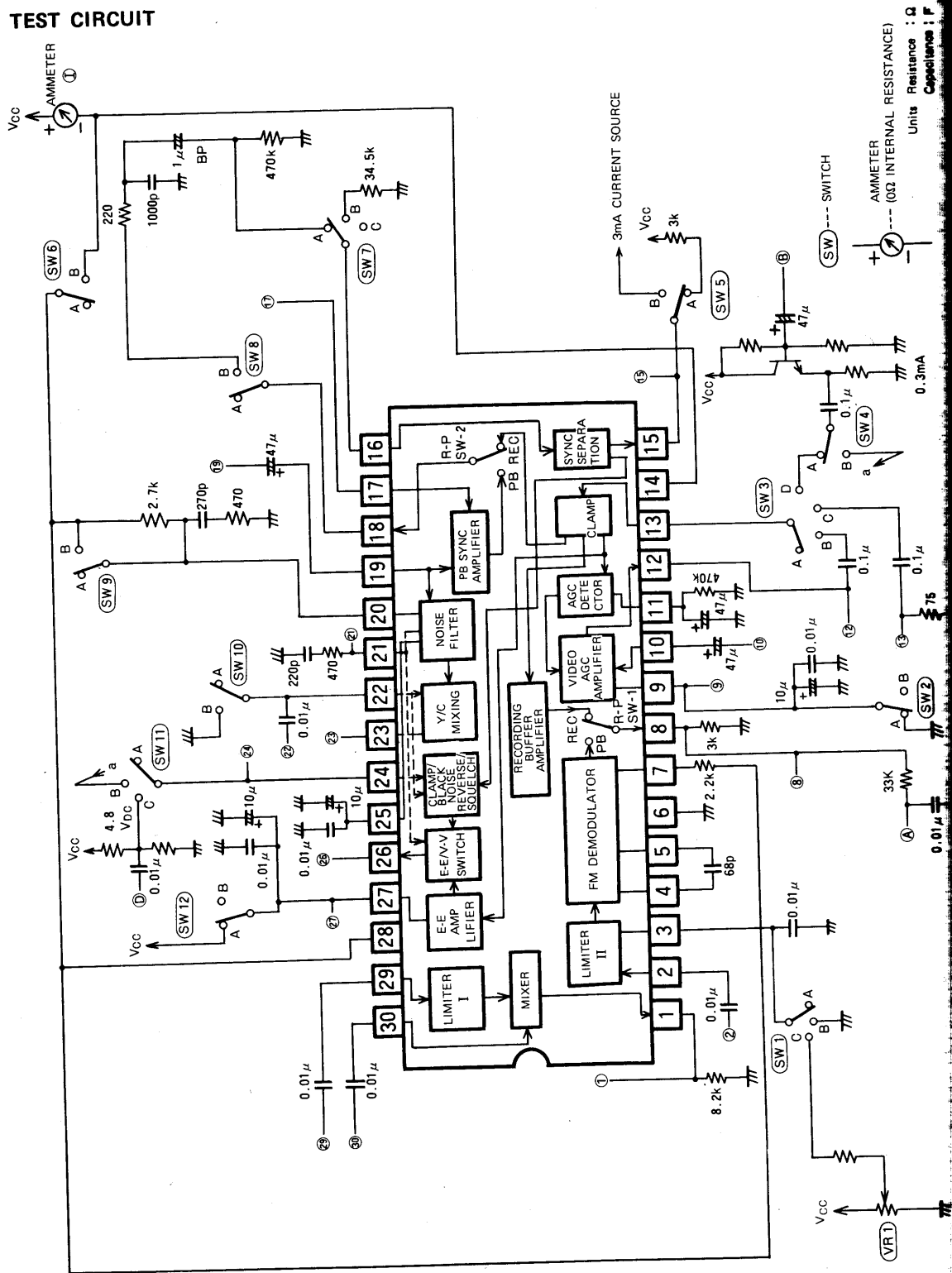
Symbol	Parameter	conditions	Limits	Unit
$V_{CC}$	Supply voltage		14	V
$P_T$	Power dissipation		1250	mW
$T_{opr}$	Operating temperature		-20 ~ +75	$^\circ\text{C}$
$T_{stg}$	Storage temperature		-40 ~ +125	$^\circ\text{C}$
$P_D$	Derating ( $T_a \geq +25^\circ\text{C}$ )		12.5	mW/ $^\circ\text{C}$

**ELECTRICAL CHARACTERISTICS** ( $T_a = 25^\circ\text{C}$ ,  $V_{CC} = 9.0\text{V}$ , unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{EC}$	Recording circuit current		24	38	52	mA
$I_E$	Playback circuit current		44	66	88	mA
$L_V$	Limiter I output amplitude	$f = 4\text{MHz}$ , $V_{in} = 800\text{mV}_{p-p}$	0.95	1.10	1.25	$V_{p-p}$
$L_V$	Limiter I gain	$f = 4\text{MHz}$ , $V_{in} = 20\text{mV}_{p-p}$	14.5	17.5	20.5	dB
$L_{2V}$	Limiter I output second harmonic distortion	$f = 4\text{MHz}$ , $V_{in} = 400\text{mV}_{p-p}$	—	—	-36	dB
$L_{1V}$	Limiter I mixing signal attenuation	$f = 1\text{MHz}$ , $V_{in} = 400\text{mV}_{p-p}$	-3	—	—	dB
$D_{EV}$	FM Demodulation sensitivity	$f = 3.4\text{MHz}$ , $4.4\text{MHz}$ , $V_{in} = 1V_{p-p}$	125	150	175	mV/MHz
$L_C$	Carrier leakage	$f = 4\text{MHz}$ , $V_{in} = 1V_{p-p}$	—	—	40	dB
$V_{in}$	Pin ② minimum input voltage	$f = 4\text{MHz}$	—	—	0.01	$V_{p-p}$
$L_S$	Playback sync amplifier gain	$f = 500\text{kHz}$ , $V_{in} = 200\text{mV}_{p-p}$	1.2	2.2	3.2	dB
$L_C$	DO Pulse high-level input voltage	High level minimum, $V = 2.4\text{V}$	0	—	5	mV
$L_C$	DO Pulse low-level input voltage	Low level maximum, $V = 0.4\text{V}$	—	—	-40	dB
$L_V$	Y Amplifier gain	$f = 500\text{kHz}$ , $V_{in} = 200\text{mV}_{p-p}$	1.2	2.2	3.2	dB
$L_N$	Noise filter characteristic 1	$f = 250\text{kHz}$ , $V_{in} = 30\text{mV}_{p-p}$	-4.5	-3.5	-2.5	dB
$L_N$	Noise filter characteristic 2	$f = 250\text{kHz}$ , $V_{in} = 300\text{mV}_{p-p}$	-1.5	-1.0	-0.5	dB
$L_N$	Noise filter characteristic 3	$f = 2\text{MHz}$ , $V_{in} = 30\text{mV}_{p-p}$	-9.6	-7.6	-5.6	dB
$L_N$	Noise filter characteristic 4	$f = 2\text{MHz}$ , $V_{in} = 300\text{mV}_{p-p}$	-1.5	-1.0	-0.5	dB
$L_V$	C Amplifier gain	$f = 3.58\text{MHz}$ , $V_{in} = 100\text{mV}_{p-p}$	4.2	5.2	6.2	dB
$L_V$	Playback video amplifier gain	$f = 500\text{kHz}$ , $V_{in} = 200\text{mV}_{p-p}$	5.4	6.4	7.4	dB
$L_V$	Playback Video amplifier input/output linearity	$V_A = 200\text{mV}_{p-p}$ , $1.4V_{p-p}$	-5	—	5	%
$L_{SC}$	Pin ③ squelch low-level voltage		—	—	1.05	V
$L_C$	Dropout white clipping level		0.52	0.57	0.62	V
$L_C$	Sync low-level output voltage	$I_{OL} = 3\text{mA}$	0	—	0.4	V
$L_C$	Sync output rising edge delay	$V_B = 280\text{mV}_{p-p}$	0	0.6	1.0	$\mu\text{s}$
$L_C$	Sync output pulse width	$V_B = 280\text{mV}_{p-p}$	4.8	5.2	5.6	$\mu\text{s}$
$L_C$	Sync separation operation 1	$V_C = 140\text{mV}_{p-p}$	8.4	—	9.0	$V_{p-p}$
$L_C$	Sync separation operation 2	$V_D = 490\text{mV}_{p-p}$	4.8	—	5.6	$\mu\text{s}$
$L_V$	Video AGC amplifier maximum gain	$f = 500\text{kHz}$ , $V_{in} = 200\text{mV}_{p-p}$	3.6	4.6	5.6	dB
$AGC-1$	AGC Characteristic 1	$V_D = 300\text{mV}_{p-p}$	0.37	0.47	0.57	$V_{p-p}$
$AGC-2$	AGC Characteristic 2	$V_D = 500\text{mV}_{p-p}$	0.60	0.70	0.80	$V_{p-p}$
$AGC-3$	AGC Characteristic 3	$V_D = 2000\text{mV}_{p-p}$	1.14	1.24	1.34	$V_{p-p}$
$L_V$	Recording buffer amplifier gain	$f = 500\text{kHz}$ , $V_{in} = 20\text{mV}_{p-p}$	7.6	9.6	11.6	dB
$L_V$	E-E Amplifier maximum gain	$f = 500\text{kHz}$ , $V_{in} = 200\text{mV}_{p-p}$	7.7	9.7	11.7	dB
$L_V$	Gain control characteristic 1	$f = 500\text{kHz}$ , $V_{in} = 200\text{mV}_{p-p}$	-10	-6	-2	dB
$L_V$	Gain control characteristic 2	$f = 500\text{kHz}$ , $V_{in} = 200\text{mV}_{p-p}$	—	—	-40	dB
$L_V$	E-E/V-V Switching operation 1	$f = 500\text{kHz}$ , $V_{in} = 200\text{mV}_{p-p}$	—	—	-40	dB
$L_V$	E-E/V-V Switching operation 2	High level minimum, $V = 7.5\text{V}$ $f = 500\text{kHz}$ , $V_{in} = 50\text{mV}_{p-p}$	—	—	-38	dB

VTR VIDEO SIGNAL PROCESSOR

TEST CIRCUIT



VTR VIDEO SIGNAL PROCESSOR

TEST METHODS (Switch is A, unless otherwise noted)

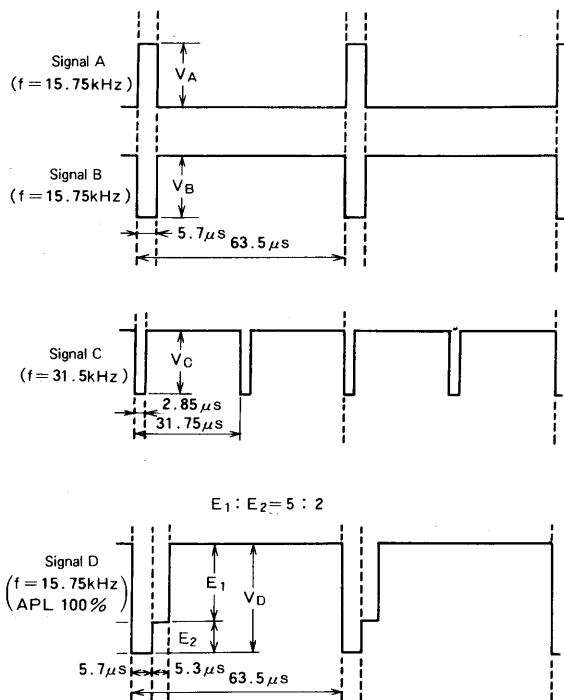
Tool	Test point	Switch conditions	Pin conditions		Test methods
			Pin	Pin conditions	
V <sub>o</sub>	①	1, 6 : B	⑳	4 MHz, 800 mV <sub>p-p</sub> input	Measure the output amplitude at point ①
V <sub>o</sub>	①	1, 6 : B	⑳	4 MHz, 20 mV <sub>p-p</sub> input	Measure the output amplitude at point ①, and determine the gain with respect to the input signal
C <sub>2IM</sub>	①	1, 6 : B	⑳	4 MHz, 400 mV <sub>p-p</sub> input	Measure the second harmonic distortion at point ①
A <sub>11</sub>	①	1, 6 : B	㉑	1 MHz, 400 mV <sub>p-p</sub> input	Measure the output amplitude at point ① and determine the attenuation with respect to the input signal.
CEV	A	6 : B	②	3.4 MHz } 1V <sub>p-p</sub> 4.4 MHz }	Measure the DC voltage at point A, with input signal of 3.4 MHz (V <sub>1</sub> ) and that of 4.4 MHz (V <sub>2</sub> ). S <sub>DEM</sub> = (V <sub>2</sub> - V <sub>1</sub> )/2
L	⑧	1 : C 6 : B	②	4 MHz, 1V <sub>p-p</sub>	Adjust VR1 so that the carrier leakage at point ⑧ is minimum. Measure the carrier leakage after adjustment.
Z <sub>1</sub>	A ②	6 : B	②	4 MHz	Reduce the input signal level at point ② to a small value and measure the input amplitude at point ② just before the DC voltage at point A changes.
S	⑬	1, 6 : B	⑬	500 kHz, 200 mV <sub>p-p</sub>	Measure the output amplitude at point ⑬ and determine the gain with respect to the input signal.
C	⑬	1, 6 : B	⑰	8 V, 2.4 V	Measure the DC voltage at point ⑬ with 8V at point ⑰ (V <sub>3</sub> ) and 2.4V at point ⑰ (V <sub>4</sub> ). V <sub>IHD</sub> =  V <sub>4</sub> - V <sub>3</sub>
L	⑬	1, 6 : B	⑰	500 kHz, 1.6 V <sub>p-p</sub> 8 V, 0.4 V	Measure the output amplitude at point ⑬ and determine the output amplitude ratio between applying 8V and 0.4V at point ⑰.
S	㉓	1, 6, 9 : B	⑰	500 kHz, 200 mV <sub>p-p</sub>	Measure the output amplitude at point ㉓ and determine the gain with respect to the input signal.
S	㉓	1, 6 : B 9 : A·B	⑰	250 kHz, 30 mV <sub>p-p</sub>	Measure the output amplitude at point ㉓, and determine the output amplitude ratio between SW9 positions A and B.
S	㉓	1, 6 : B 9 : A·B	⑰	250 kHz, 300 mV <sub>p-p</sub>	
S	㉓	1, 6 : B 9 : A·B	⑰	2 MHz, 30 mV <sub>p-p</sub>	
S	㉓	1, 6 : B 9 : A·B	⑰	2 MHz, 300 mV <sub>p-p</sub>	
S	㉓	1, 6 : B	㉒	3.58 MHz, 100 mV <sub>p-p</sub>	Measure the output amplitude at point ㉓, and determine the gain with respect to the input signal.
S	㉔	1, 6 : B 11 : C	㉔	500 kHz, 200 mV <sub>p-p</sub>	Measure the output amplitude at point ㉔, and determine the gain with respect to the input signal.
S	㉔	1, 4, 6, 8, 11 : B	⑰ ㉔	Signal B : 200 mV <sub>p-p</sub> Signal A : 200 mV <sub>p-p</sub> , 1.4 V <sub>p-p</sub>	Measure the output amplitude at point ㉔, with input signal A of 200 mV <sub>p-p</sub> (V <sub>5</sub> ), and that of 1.4V <sub>p-p</sub> (V <sub>6</sub> ). L <sub>PV</sub> = (V <sub>6</sub> /7 × V <sub>5</sub> ) × 100 (%)
L <sub>SO</sub>	㉔	1, 6 : B	㉔ C	V <sub>24 SYNC</sub> - 1.4 V Applied DC voltage	With switch SW7 in the B position, and point ㉔ open, let the DC voltages at point ㉔ and point ㉔ be V <sub>24 SYNC</sub> and V <sub>26 SYNC</sub> , respectively. With switch SW7 in the A position, apply V <sub>24 SYNC</sub> - 1.4V to point ㉔, and gradually decrease the voltage applied to point C from 2V. Determine the voltage applied to point C at the time that the DC voltage at point ㉔ just falls below V <sub>26 SYNC</sub> .
C <sub>L</sub>	㉔	1, 6 : B	㉔	V <sub>24 SYNC</sub> + 0.5 V	Measure the DC voltage at point ㉔ and let this be V <sub>7</sub> . WCL = V <sub>7</sub> - V <sub>26 SYNC</sub> /G <sub>PV</sub>
S <sub>L</sub>	⑮	5 : B 7 : C	⑮	3 mA	Measure the DC voltage at point ⑮
S <sub>C</sub>	B ⑮	3 : D 8 : B	B	Signal B : 280 mV <sub>p-p</sub>	Measure the pulse delay time from the point B to the point ⑮
S <sub>A</sub>	⑮	3 : D 8 : B	B	Signal B : 280 mV <sub>p-p</sub>	Measure the pulse width at point ⑮
S <sub>1</sub>	⑮	3 : D 8 : B	B	Signal C : 140 mV <sub>p-p</sub>	Measure the output amplitude at point ⑮
S <sub>2</sub>	⑮	3 : D 8 : B	B	Signal D : 490 mV <sub>p-p</sub>	Measure the pulse width at point ⑮
S <sub>4</sub>	⑫		⑩	500 kHz, 200 mV <sub>p-p</sub>	Measure the output amplitude at point ⑫, and determine the gain with respect to the input signal.
AGC1 AGC2 AGC3	⑫	2, 3 : B	⑨ ⑩	Applied DC voltage Signal D : 300 mV <sub>p-p</sub> (V <sub>AGC1</sub> ) 500 mV <sub>p-p</sub> (V <sub>AGC2</sub> ) 2000 mV <sub>p-p</sub> (V <sub>AGC3</sub> )	With a signal D of 1V <sub>p-p</sub> input to point ⑩, adjust the DC voltage applied to point ⑨ so that the output amplitude at point ⑫ is 1V <sub>p-p</sub> (V <sub>P9</sub> ). Measure the output amplitude at point ⑫ with respect to the various input signals.
EC	⑧	3 : C	⑬	500 kHz, 200 mV <sub>p-p</sub>	Measure the output amplitude at point ⑧, and determine the gain with respect to the input signal.

VTR VIDEO SIGNAL PROCESSOR

TEST METHODS (Switch is A, unless otherwise noted)

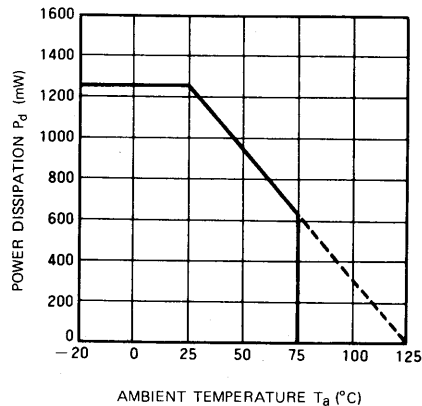
Symbol	Test point	Switch conditions	Pin	Pin conditions	Test method
G <sub>EE</sub>	⑳	3 : C	⑬ ⑳	500 kHz, 200 mV <sub>p-p</sub> 9 V	Measure the output amplitude at point ⑳, and determine the gain with respect to the input signal.
G <sub>C1</sub>	⑳	3 : C 12 : B	⑬ ㉑	500 kHz, 200 mV <sub>p-p</sub> 4.1 V	Measure the output amplitude at point ⑳ and determine the gain with respect to the input signal.
G <sub>C2</sub>	⑳	3 : C 12 : B	⑬ ㉑	500 kHz, 200 mV <sub>p-p</sub> 1.8 V	Measure the output amplitude at point ⑳, and determine the gain with respect to the input signal.
E <sub>V1</sub>	⑳	3 : C 7 : B	⑬ ㉑	500 kHz, 500 mV <sub>p-p</sub> DC Voltage applied	Measure the output amplitude at point ⑳, and determine the output amplitude ratio between the output when DC Voltage (point ㉑ open + 1V) is applied to point ㉑ and that when 9V is applied to point ㉑.
E <sub>V2</sub>	⑳	1, 6 : B 11 : C	① ㉑	500 kHz, 500 mV <sub>p-p</sub> Open, 7.5V	Measure the output amplitude at point ⑳, and determine the output amplitude ratio of the output when point ㉑ is open to that when 7.5V is applied to point ㉑.
I <sub>REC</sub>	①		㉑	9 V	Measure the current at ammeter ①
I <sub>PB</sub>	①	6 : B			Measure the current at ammeter ①

Waveforms for Signal A, B, C, and D



TYPICAL CHARACTERISTICS

THERMAL DERATING (MAXIMUM RATINGS)

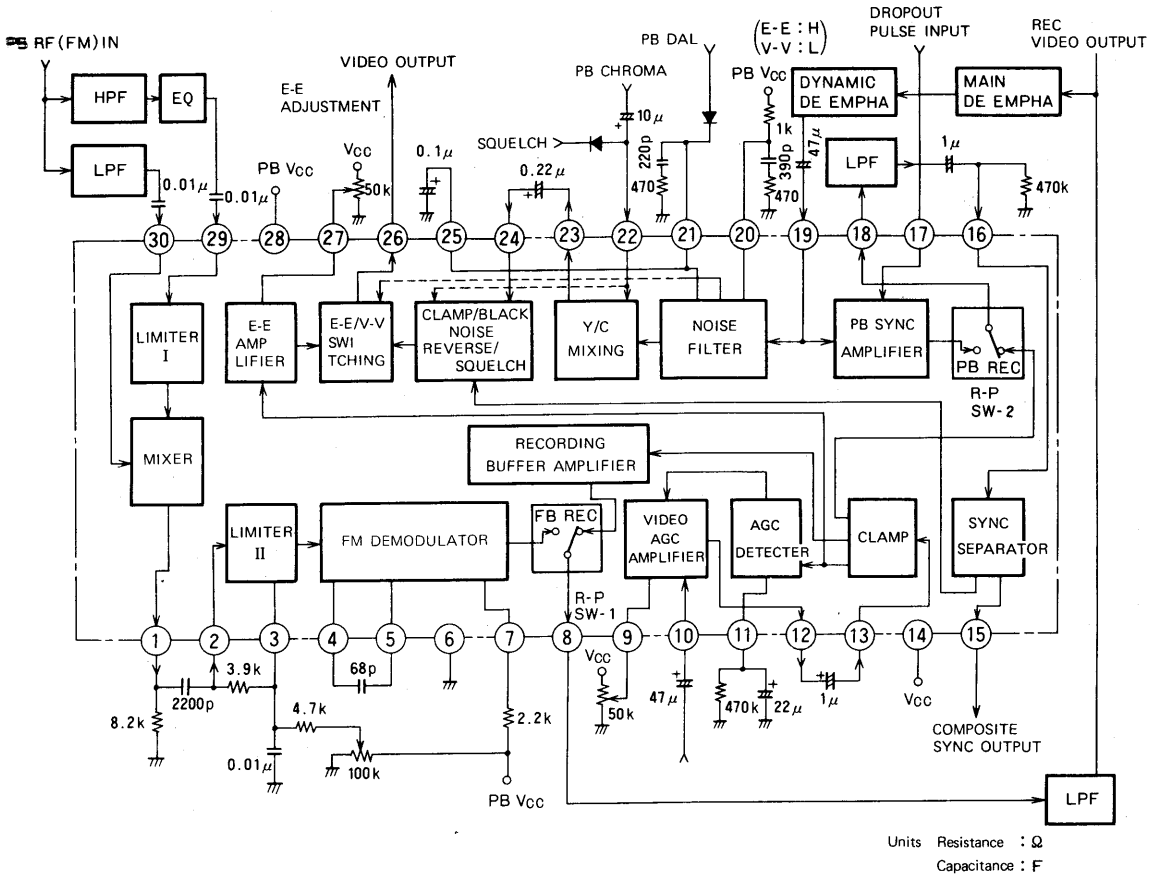


Standard Input Signal Levels

Pin	Signal	Standard level
②	RF (FM)	1.05V <sub>p-p</sub>
⑩	E-E Video	1.0V <sub>p-p</sub>
⑬	E-E Video	1.0V <sub>p-p</sub>
⑰	PB Y-Video	0.74V <sub>p-p</sub>
㉒	PB Chroma	0.32V <sub>p-p</sub>
㉔	PB Video	0.96V <sub>p-p</sub>
㉖	RF (FM) ..... high frequency	0.4V <sub>p-p</sub>
㉘	RF (FM) ..... low frequency	0.4V <sub>p-p</sub>

**VTR VIDEO SIGNAL PROCESSOR**

**APPLICATION EXAMPLE**



# MITSUBISHI LINEAR ICs

## M51451AP

### VTR FM SIGNAL PROCESSOR

#### DESCRIPTION

The M51451AP is a semiconductor integrated circuit designed for use in VTR FM signal processing. It consists of pre-amplifiers (channel 1 and channel 2), FM AGC circuit, dropout compensator, sync clamp circuit, non-linear pre-emphasis circuit, main pre-emphasis circuit, white/dark clipping circuit, and FM modulator.

#### FEATURES

- Packaged in a 30-pin DIL shrink package to minimize required PC board space
- Uses a low-noise, high-gain preamplifier circuit
- Built-in non-linear pre-emphasis circuit
- Used in conjunction with the M51450P, all major VTR video signal processing functions may be performed.

#### PIN CONFIGURATION (TOP VIEW)

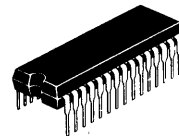
DOC TIME CONSTANT	1	30	GND
PLAYBACK CIRCUIT SUPPLY VOLTAGE (9V)	2	29	DROPOUT PULSE OUTPUT
FM AGC BYPASS CAPACITOR	3	28	PREAMP CHANNEL SWITCHING PULSE INPUT
FM AGC INPUT	4	27	GND
FM AGC TIME CONSTANT	5	26	PREAMP (CH-1) FEEDBACK CAPACITOR
DOC 1H DELAY SIGNAL INPUT	6	25	PREAMP (CH-1) INPUT
FM AGC/DOC OUTPUT	7	24	PREAMP GND
FM MODULATOR OSCILLATION INHIBIT PULSE INPUT	8	23	PREAMP (CH-2) INPUT
PREAMP/PB/FM MODULATOR (RECORDING) OUTPUT	9	22	PREAMP (CH-2) FEEDBACK CAPACITOR
RECORDING CIRCUIT SUPPLY VOLTAGE (9V)	10	21	PREAMP SUPPLY VOLTAGE (9V)
FM MODULATOR CAPACITOR	11	20	SYNC SIGNAL INPUT
FM MODULATOR FREQUENCY ADJUSTMENT RESISTOR	12	19	RECORDING VIDEO SIGNAL INPUT
DARK CLIPPING LEVEL SETTING	13	18	NON-LINEAR PRE-EMPHASIS TIME CONSTANT
WHITE CLIPPING LEVEL SETTING	14	17	MAIN PRE-EMPHASIS TIME CONSTANT
	15	16	MAIN PRE-EMPHASIS FEEDBACK RESISTOR

#### APPLICATION

VTR FM signal processing

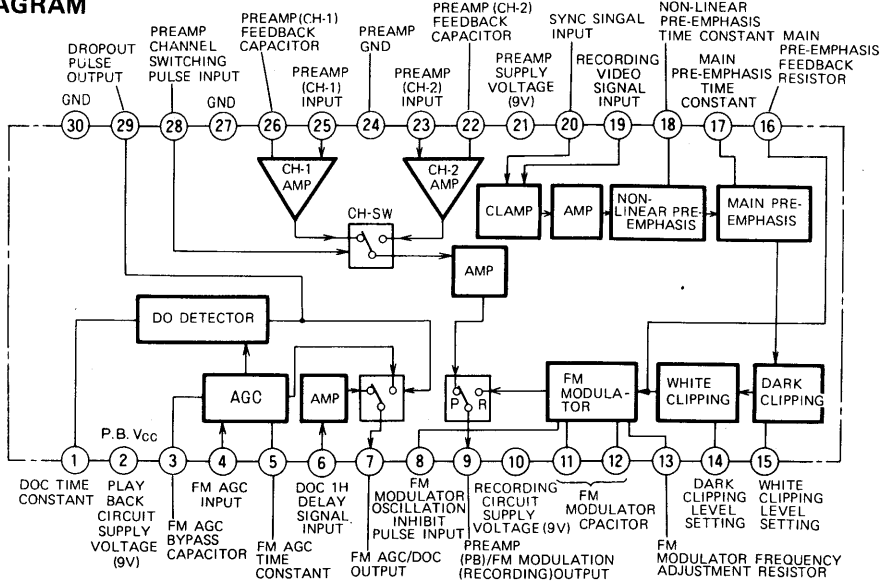
#### RECOMMENDED OPERATING CONDITIONS

Supply voltage range ..... 8~10V  
 Rated supply voltage ..... 9V



30-pin molded plastic DIL (shrink)

#### BLOCK DIAGRAM



**VTR FM SIGNAL PROCESSOR**

**ABSOLUTE MAXIMUM RATINGS** (  $T_a = 25^\circ\text{C}$ , unless otherwise noted)

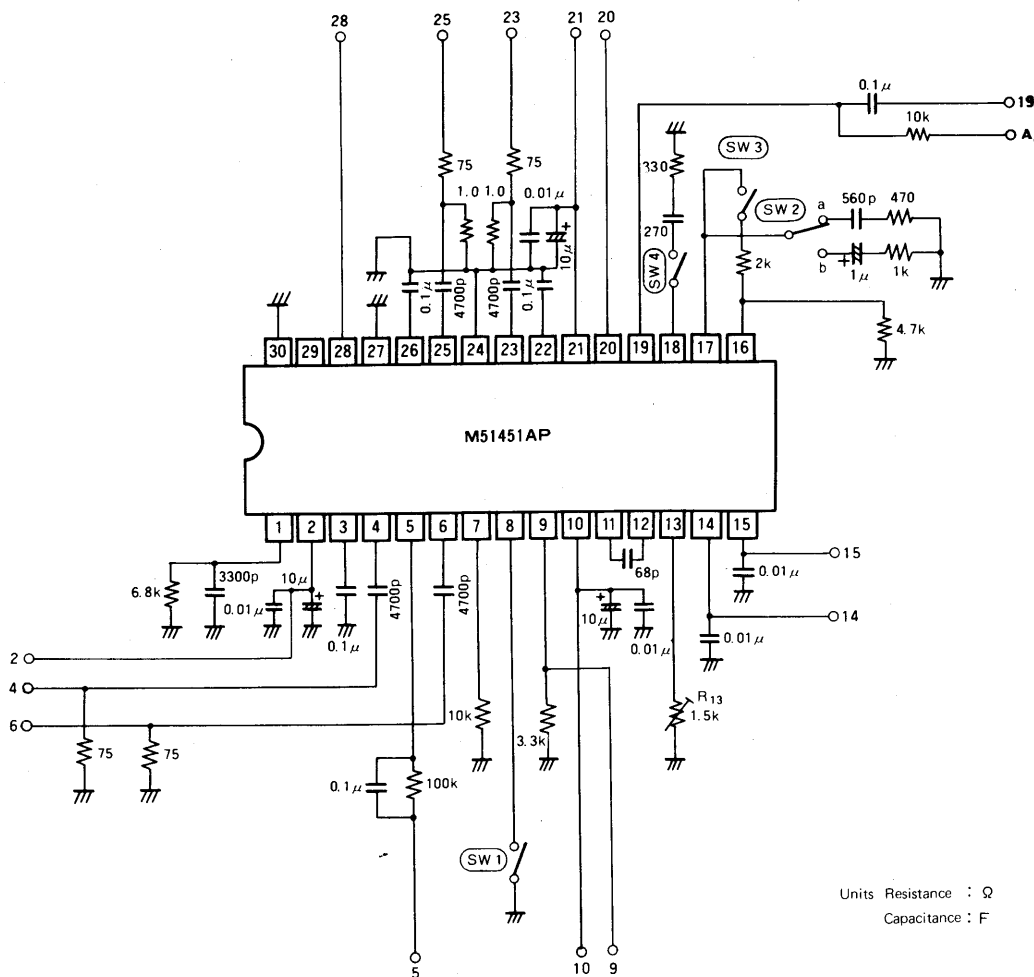
Symbol	Parameter	Conditions	Limits	Unit
$V_{CC}$	Supply voltage		14	V
$P_T$	Power dissipation		1250	mW
$T_{OC}$	Operating temperature		-20 ~ +75	$^\circ\text{C}$
$T_{SG}$	Storage temperature		-40 ~ +125	$^\circ\text{C}$
$I_P$	Derating ( $T_a \geq +25^\circ\text{C}$ )		12.5	mW/deg

**ELECTRICAL CHARACTERISTICS** (  $T_a = 25^\circ\text{C}$ ,  $V_{CC} = 9\text{V}$ , unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
$I_{CA}$	Preamplifier circuit current		3.7	5.5	7.3	mA
$I_{CB}$	Playback circuit current		14.8	19.8	24.8	mA
$I_{CC}$	Recording circuit current		12.6	18.0	23.4	mA
$G_{PA1}$	Preamplifier (CH-1) voltage gain	$f = 4\text{MHz}$ , $V_{IN} = 400\mu\text{V}_{P-P}$	57.3	60.3	63.3	dB
$G_{PA2}$	Preamplifier (CH-2) voltage gain	$f = 4\text{MHz}$ , $V_{IN} = 400\mu\text{V}_{P-P}$	57.3	60.3	63.3	dB
$G_{PC1}$	Preamp channel-to-channel voltage gain difference	$f = 4\text{MHz}$ , $V_{IN} = 400\mu\text{V}_{P-P}$	-2	—	2	dB
$A_{NPA1}$	Preamplifier (CH-1) noise voltage	$f_N = 1\text{MHz}$ , $\Delta f_N = 120\text{kHz}$	—	—	2	$\mu\text{V}$
$A_{NPA2}$	Preamplifier (CH-2) noise voltage	$f_N = 1\text{MHz}$ , $\Delta f_N = 120\text{kHz}$	—	—	2	$\mu\text{V}$
$G_{AGC}$	AGC Amplifier voltage gain	$f = 4\text{MHz}$ , $V_{IN} = 50\text{mV}_{P-P}$	11.3	14.3	17.3	dB
$V_{O-AGC}$	AGC Output voltage	$f = 4\text{MHz}$ , $V_{IN} = 400\text{mV}_{P-P}$	300	600	900	$\text{mV}_{P-P}$
$DC_{DET}$	Dropout detection level	$f = 4\text{MHz}$ , $0\text{dB}$ : $V_O = 600\text{mV}_{P-P}$	-25	-22	-19	dB
$DC_{HYS}$	DOC Hysteresis	$f = 4\text{MHz}$ , with the dropout detector output at 0 dB.	1.5	4.0	6.5	dB
$G_{CLA}$	Delayed signal amplifier voltage gain	$f = 4\text{MHz}$ , $V_{IN} = 90\text{mV}_{P-P}$	11.8	14.8	17.8	dB
$A_{V-REC}$	Recording video signal input amplitude		210	280	350	mV
$V_{FM-DO}$	FM Modulator output voltage	$f = 4\text{MHz}$	0.8	1.0	1.2	$\text{V}_{P-P}$
$G_{NL1}$	Non-linear pre-emphasis characteristic 1	Ratio of outputs at 10 kHz and 500 kHz $V_{IN} = 30\text{mV}_{P-P}$	4.5	6.0	7.5	dB
$G_{NL2}$	Non-linear pre-emphasis characteristic 2	Ratio of outputs at 10 kHz and 2 MHz $V_{IN} = 280\text{mV}_{P-P}$	0	2.0	4.0	dB
$G_{NL3}$	Non-linear pre-emphasis characteristic 3	Ratio of outputs at 10 kHz and 2 MHz $V_{IN} = 90\text{mV}_{P-P}$	2.5	4.5	6.5	dB
$G_{NL4}$	Non-linear pre-emphasis characteristic 4	Ratio of outputs at 10 kHz and 2 MHz $V_{IN} = 30\text{mV}_{P-P}$	4.5	6.5	8.5	dB
$G_{MPE1}$	Main pre-emphasis characteristic 1	Ratio of outputs at 10 kHz and 500 kHz $V_{IN} = 50\text{mV}_{P-P}$	9.3	10.3	11.3	dB
$G_{MPE2}$	Main pre-emphasis characteristic 2	Ratio of outputs at 10 kHz and 3 MHz $V_{IN} = 50\text{mV}_{P-P}$	12.3	13.8	15.3	dB



TEST CIRCUIT



**VTR FM SIGNAL PROCESSOR**

**TEST METHODS**

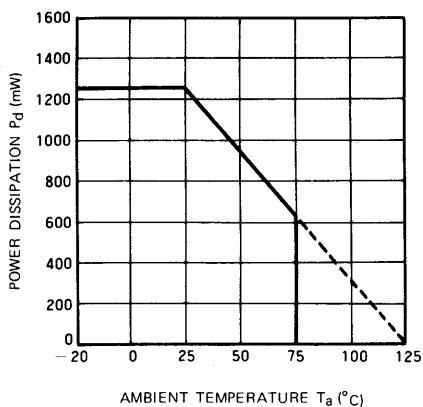
Symbol	Switch settings	Pin conditions														Test method											
		2	4	5	6	10	14	15	19	20	21	23	25	28	A												
$I_{PA}$		9V		9V												9V										Measure current at pin ⑫	
$I_{PB}$		9V		9V												9V										Measure current at pin ⑫	
$I_{REC}$	1,3 : ON									9V	1V	8V				9V										Adjust the value of $R_{13}$ so that the output at pin ⑨ is 3.4 MHz, measure the current at pin ⑩ and the voltage at A (the DC voltage at A is $V_{19 SYNC}$ )	
$G_{PA1}$		9V		9V												9V	0V	4MHz 30.4 mVp-p	0V							Measure the output signal amplitude at pin ⑦. Determine the gain with respect to the input signal (400 $\mu$ Vp-p)	
$G_{PA2}$		9V		9V												9V	4MHz 30.4 mVp-p	0V	9V							Measure the output signal amplitude at pin ⑦. Determine the gain with respect to the input signal (400 $\mu$ Vp-p)	
$G_{DCA}$																										$GDPA = G_{PA2} - G_{PA1}$	
$V_{FPA1}$		9V		9V												9V	0V	0V	0V							Connect a filter with an equivalent noise bandwidth of 120 kHz to pin ⑨ and measure the rms noise voltage at the filter output.	
$V_{FPA2}$		9V		9V												9V	0V	0V	9V							Connect a filter with an equivalent noise bandwidth of 120 kHz to pin ⑨ and measure the rms noise voltage at the filter output.	
$V_{DOC}$		9V	4MHz 50mVp-p	9V												9V										Measure the output signal amplitude at pin ⑦. Determine the gain with respect to the input signal.	
$V_{D13C}$		9V	4MHz 400mVp-p	9V												9V										Measure the output signal amplitude at pin ⑦.	
$DOC_{DET}$		9V	4MHz	9V												9V										Decrease the input signal at pin ④ gradually from 400mVp-p and measure the output signal amplitude at pin ⑦ at the point at which pin ② (open-collector SW output) turns on. Ratio of the reference output level of 600mVp-p.	
$DOC_{LVS}$		9V	4MHz	9V												9V										Increase the input signal level at pin ④ gradually from 0 and measure the output signal amplitude at pin ⑦ at the point at which pin ② (open-collector SW output) turns off, determining the ratio of this level to the $DOC_{det}$ amplitude.	
$G_{PA}$		9V	0V	9V	4MHz 90mVp-p											9V										Measure the output signal amplitude at pin ⑦. Determine the gain with respect to the input signal.	
$V_{FVDO}$	1,3 : ON									9V	1V	8V				0V								Applied DC Voltage		Gradually increase the applied DC voltage at A from $V_{19 SYNC}$ . Measure the output signal amplitude when the output signal frequency at pin ⑨ is 4.0 MHz.	
$V_{FREC}$																										$f_{SYNC} = 3.4 \text{ MHz}$ $f_{WHITE} = 4.4 \text{ MHz}$	
$V_{FE1}$	1: OFF 2: b 3,4 : ON									9V	1V	8V	10kHz 500kHz 30mVp-p			0V										$V_{19 SYNC} + 0.14V$	Measure the output signal amplitude at pin ⑬. Determine the output ratios for 500 kHz, 2 MHz, and 3 MHz, with respect to an input signal of 10 kHz.
$V_{FE2}$										9V	1V	8V	10kHz 2MHz 280mVp-p			0V										$V_{19 SYNC} + 0.14V$	
$V_{FE3}$										9V	1V	8V	10kHz 2MHz 90mVp-p			0V										$V_{19 SYNC} + 0.14V$	
$V_{FE4}$										9V	1V	8V	10kHz 2MHz 30mVp-p			0V										$V_{19 SYNC} + 0.14V$	

TEST METHODS

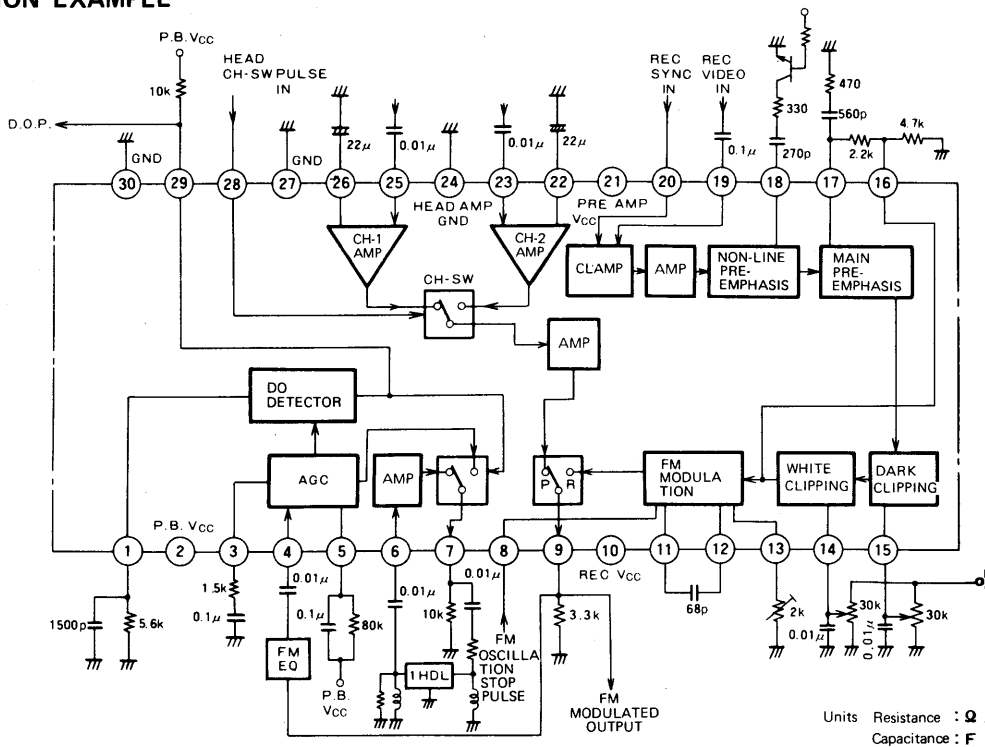
Symbol	Switch settings	Pin conditions														Test method	
		2	4	5	6	10	14	15	19	20	21	23	25	28	A		
GMPE1	1,4 : OFF 2 : a 3 : ON					9V	1V	8V	10kHz 500kHz 50mV <sub>P-P</sub>	0V						V <sub>19 SYNC</sub> +0.14V	Measure the output signal amplitude at pin ⑬. Determine the output ratios for 500 kHz, 2 MHz, and 3 MHz, with respect to an input signal of 10 kHz.
GMPE2					9V	1V	8V	10kHz 3MHz 50mV <sub>P-P</sub>	0V						V <sub>19 SYNC</sub> +0.14V		

TYPICAL CHARACTERISTICS

THERMAL DERATING (MAXIMUM RATING)



APPLICATION EXAMPLE



# M51453P

## VIDEO SIGNAL NOISE CANCELER

### DESCRIPTION

The M51453P is a semiconductor integrated circuit designed to cancel out the noise in VTR video signals. It consists of a limiter, FM demodulator, comparator and clipper.

### FEATURES

- Main video input pins with dual polarity
- Detected noise output provided with dual polarity
- Great improvement in S/N ratio (3dB typ)

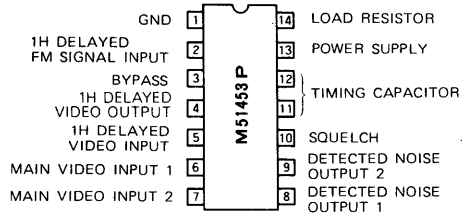
### APPLICATION

VTR video signal noise cancellation

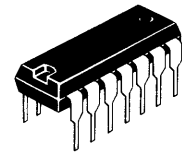
### RECOMMENDED OPERATING CONDITIONS

Supply voltage range . . . . . 8~10V  
 Rated supply voltage . . . . . 9V

### PIN CONFIGURATION (TOP VIEW)

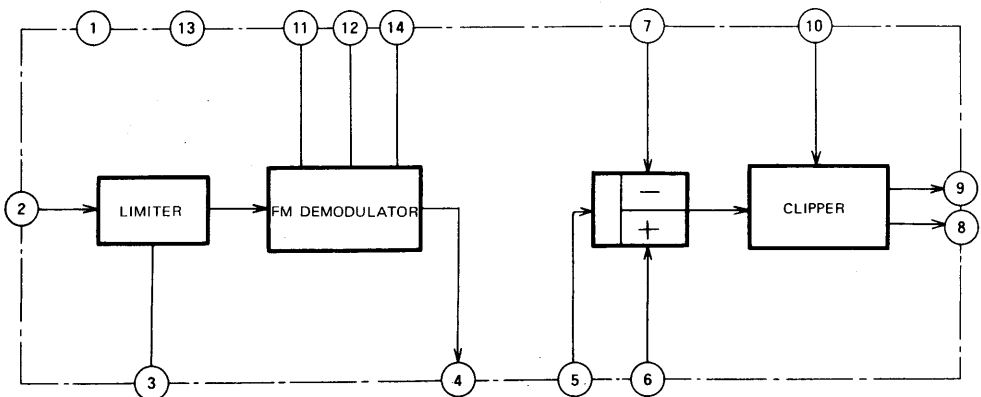


NC: NO CONNECTION



14-pin molded plastic DIL

### BLOCK DIAGRAM



VIDEO SIGNAL NOISE CANCELER

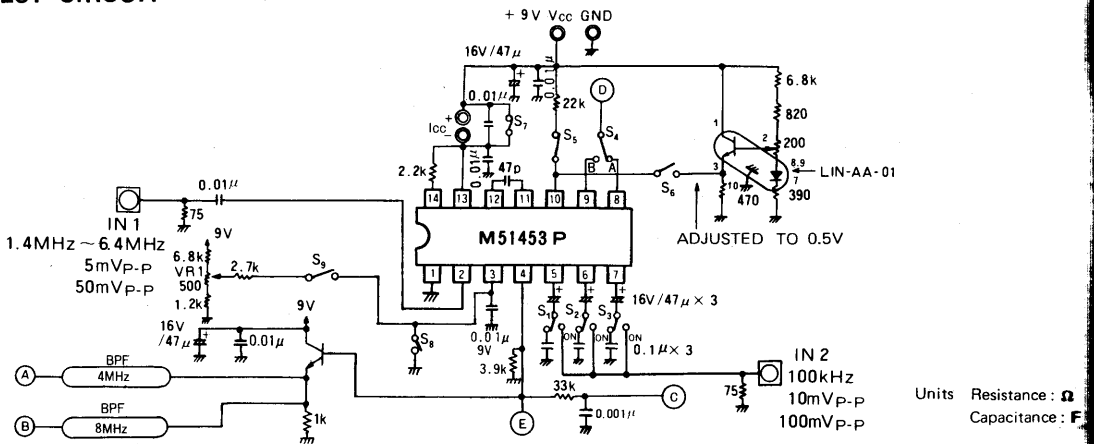
ABSOLUTE MAXIMUM RATINGS (T<sub>a</sub>=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Limits	Unit
V <sub>CC</sub>	Supply voltage		14	V
P <sub>d</sub>	Power dissipation		950	mW
T <sub>opg</sub>	Ambient temperature		-20 ~ +75	°C
T <sub>stg</sub>	Storage temperature		-40 ~ +125	°C
K <sub>θ</sub>	Thermal derating (T <sub>a</sub> ≥ +25°C)		9.5	mW/°C

ELECTRICAL CHARACTERISTICS (T<sub>a</sub>=25°C, V<sub>CC</sub>=9V, unless otherwise noted)

Symbol	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
S <sub>DEM</sub>	FM demodulation sensitivity	Pin ③ input level 100mV <sub>p-p</sub> f=3.4MHz, 4.4MHz	95	115	135	mV/MHz
L <sub>DEM-1</sub>	FM demodulation linearity 1	f=6.4MHz	-5	0	5	%
L <sub>DEM-2</sub>	FM demodulation linearity 2	f=1.4MHz	-5	0	5	%
CL	Carrier leakage	f=4MHz			-40	dB
S <sub>RF</sub>	RF sensitivity	Potential difference at ② point between pin ② input 100mV <sub>p-p</sub> and 10mV <sub>p-p</sub> (f=4MHz)	-5	0	5	mV
G <sub>V-1</sub>	Gain 1	Pin ⑤ input 10mV <sub>p-p</sub> (100kHz)	13	16	19	dB
G <sub>V-2</sub>	Gain 2	Pin ⑥ input 10mV <sub>p-p</sub> (100kHz)	13	16	19	dB
G <sub>V-3</sub>	Gain 3	Pin ⑦ input 10mV <sub>p-p</sub> (100kHz)	13	16	19	dB
R <sub>CLP</sub>	Clipping capacity	Output ratio between pin ⑤ input 50mV <sub>p-p</sub> and 100mV <sub>p-p</sub> (100kHz)	75	85	95	%
G <sub>OFS-1</sub>	Gain offset 1	100mV <sub>p-p</sub> (100kHz) input simultaneously at pins ⑤ and ⑦			50	mV <sub>p-p</sub>
G <sub>OFS-2</sub>	Gain offset 2	100mV <sub>p-p</sub> (100kHz) input simultaneously at pin ⑥ and ⑦			50	mV <sub>p-p</sub>
SQ	Squelch capacity	Pin ⑤ input 100mV <sub>p-p</sub> (100kHz), +0.5V at pin ⑩			10	mV <sub>p-p</sub>
G <sub>F-1</sub>	Frequency response 1	Output ratio between pin ⑤ input 10mV <sub>p-p</sub> , 100kHz and 6MHz	-2		0	dB
G <sub>F-2</sub>	Frequency response 2	Output ratio between pin ⑥ input 10mV <sub>p-p</sub> , 100kHz and 6MHz	-2		0	dB
G <sub>F-3</sub>	Frequency response 3	Output ratio between pin ⑦ input 10mV <sub>p-p</sub> , 100kHz and 6MHz	-2		0	dB
I <sub>CC-1</sub>	Supply current 1	Pin ⑩ open	9.3	12.5	15.6	mA
I <sub>CC-2</sub>	Supply current 2		10.8	14.4	18	mA

TEST CIRCUIT

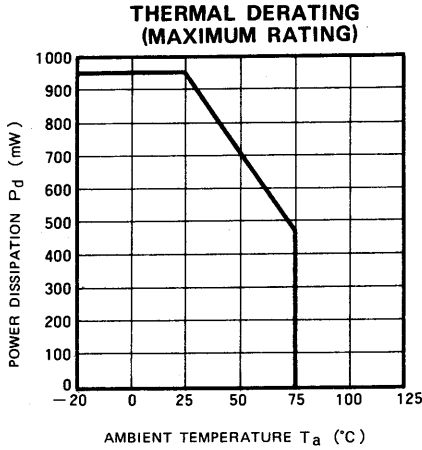


VIDEO SIGNAL NOISE CANCELER

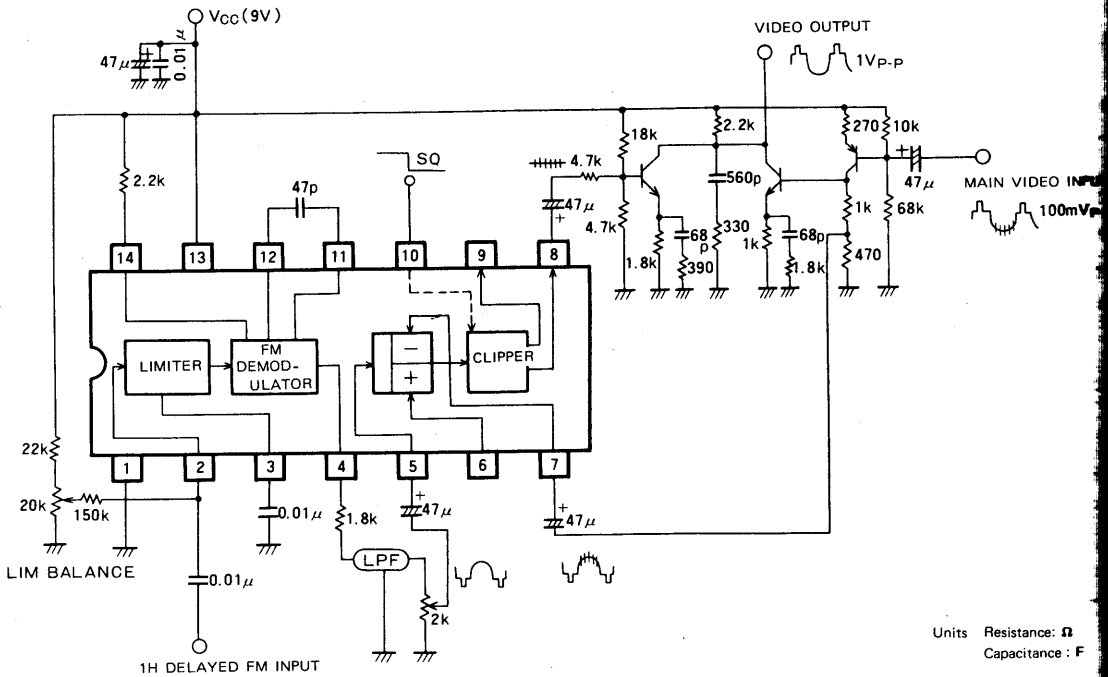
TEST METHODS (V<sub>CC</sub> = 9V, T<sub>a</sub> = 25°C, unless otherwise noted)

Symbol	S <sub>1</sub>	S <sub>2</sub>	S <sub>3</sub>	S <sub>4</sub>	S <sub>5</sub>	S <sub>6</sub>	S <sub>7</sub>	S <sub>8</sub>	S <sub>9</sub>	Method
S <sub>DEV</sub>	OFF	OFF	OFF	—	ON	OFF	ON	OFF	OFF	If pin ② 3.4MHz, 4.4MHz, 100mV <sub>p-p</sub> input, point ③ output voltages are made V <sub>1</sub> and V <sub>2</sub> , respectively: S <sub>DEM</sub> = (V <sub>2</sub> - V <sub>1</sub> ) × 1000/2 (mV/MHz)
L <sub>DEM-1</sub>	OFF	OFF	OFF	—	ON	OFF	ON	OFF	OFF	If pin ② 6.4MHz, 100V <sub>p-p</sub> input, point ③ output voltage is made V <sub>3</sub> : L <sub>DEM-1</sub> = $\left\{ \frac{V_3 - V_2}{2(V_2 - V_1)} - 1 \right\} \times 100$ (%)
L <sub>DEM-2</sub>	OFF	OFF	OFF	—	ON	OFF	ON	OFF	OFF	If pin ② 1.4MHz, 100mV <sub>p-p</sub> input, point ③ output voltage is made V <sub>4</sub> : L <sub>DEM-2</sub> = $\left\{ \frac{V_1 - V_4}{2(V_2 - V_1)} - 1 \right\} \times 100$ (%)
S <sub>CL</sub>	OFF	OFF	OFF	—	ON	OFF	ON	OFF	ON	Pin ② 4 MHz, 100mV <sub>p-p</sub> (less than -50dB second harmonic level) input, point ④ pin ④ observed on spectrum analyzer, VR1 adjusted for minimum 4MHz component level, 8MHz: 4MHz component ratio read in decibels
S <sub>RF</sub>	OFF	OFF	OFF	—	ON	OFF	ON	OFF	OFF	If pin ② 4MHz, 100mV <sub>p-p</sub> , 10mV <sub>p-p</sub> input, point ③ output voltages are made V <sub>7</sub> and V <sub>8</sub> , respectively: S <sub>RF</sub> = (V <sub>7</sub> - V <sub>8</sub> ) × 1000 (mV)
G <sub>V-1</sub>	ON	OFF	OFF	A B	ON	OFF	ON	ON	OFF	If pin ⑤ 100kHz, 10mV <sub>p-p</sub> input, point ④ signal voltage is made V <sub>2</sub> (mV <sub>p-p</sub> ): G <sub>V-1</sub> = 20 log <sub>10</sub> $\frac{V_2}{10}$ (dB)
G <sub>V-2</sub>	OFF	ON	OFF	A B	ON	OFF	ON	ON	OFF	If pin ⑥ 100kHz, 10mV <sub>p-p</sub> input, point ④ signal voltage is made V <sub>2</sub> (mV <sub>p-p</sub> ): G <sub>V-2</sub> = 20 log <sub>10</sub> $\frac{V_2}{10}$ (dB)
G <sub>V-3</sub>	OFF	OFF	ON	A B	ON	OFF	ON	ON	OFF	If pin ⑦ 100kHz, 10mV <sub>p-p</sub> input, point ④ signal voltage is made V <sub>3</sub> (mV <sub>p-p</sub> ): G <sub>V-3</sub> = 20 log <sub>10</sub> $\frac{V_3}{10}$ (dB)
R <sub>CLP</sub>	ON	OFF	OFF	A B	ON	OFF	ON	ON	OFF	If pin ⑤ 100kHz, 50mV <sub>p-p</sub> , 100mV <sub>p-p</sub> input, point ④ output voltages are made V <sub>4</sub> and V <sub>5</sub> , respectively: R <sub>CLP</sub> = $\frac{V_4}{V_5} \times 100$ (%)
S <sub>RF-1</sub>	ON	OFF	ON	A B	ON	OFF	ON	ON	OFF	Simultaneous pin ⑤ and pin ⑦ 100kHz, 100mV <sub>p-p</sub> input, point ④ output voltage measured
S <sub>RF-2</sub>	OFF	ON	ON	A B	ON	OFF	ON	ON	OFF	Simultaneous pin ⑥ and pin ⑦ 100kHz, 100mV <sub>p-p</sub> input, point ④ output voltage measured
S <sub>DC</sub>	ON	OFF	OFF	A B	OFF	ON	ON	ON	OFF	If pin ⑤ 100kHz, 100mV <sub>p-p</sub> input, DC + 0.5V applied to pin ⑩, point ④ output voltage measured
G <sub>F-1</sub>	ON	OFF	OFF	A B	ON	OFF	ON	ON	OFF	If pin ⑤ 100kHz, 6MHz, 10mV <sub>p-p</sub> input, point ④ output voltages are made V <sub>6</sub> and V <sub>7</sub> , respectively: G <sub>F-1</sub> = 20 log <sub>10</sub> $\frac{V_7}{V_6}$ (dB)
G <sub>F-2</sub>	OFF	ON	OFF	A B	ON	OFF	ON	ON	OFF	If pin ⑥ 100kHz, 6MHz, 10mV <sub>p-p</sub> input, point ④ output voltages are made V <sub>8</sub> and V <sub>9</sub> , respectively: G <sub>F-2</sub> = 20 log <sub>11</sub> $\frac{V_9}{V_8}$ (dB)
G <sub>F-3</sub>	OFF	OFF	ON	A B	ON	OFF	ON	ON	OFF	If pin ⑦ 100kHz, 6MHz, 10mV <sub>p-p</sub> input, point ④ output voltages are made V <sub>10</sub> and V <sub>11</sub> respectively: G <sub>F-3</sub> = 20 log <sub>10</sub> $\frac{V_{11}}{V_{10}}$ (dB)
S <sub>IO</sub>	OFF	OFF	OFF	—	OFF	OFF	OFF	OFF	OFF	Pin ⑩ open, pin ③ current measured
S <sub>IO-2</sub>	OFF	OFF	OFF	—	ON	OFF	OFF	OFF	OFF	22kΩ connected across pin ⑩ and V <sub>CC</sub>

TYPICAL CHARACTERISTICS (T<sub>a</sub> = 25°C, unless otherwise noted)



APPLICATION EXAMPLE



# M51454L

## VIDEO SIGNAL NOISE CANCELER

### DESCRIPTION

M51454L is a semiconductor integrated circuit designed to cancel out the noise in VTR video signals. It consists of a limiter, FM demodulator, comparator, clipper and adder.

### FEATURES

- Housed in a 16-pin ZIL package and so takes up very little room on the pcb
- Limiter balance adjustments not required
- Minimal number of external parts
- Great improvement in S/N ratio (3dB typ)

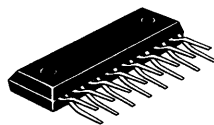
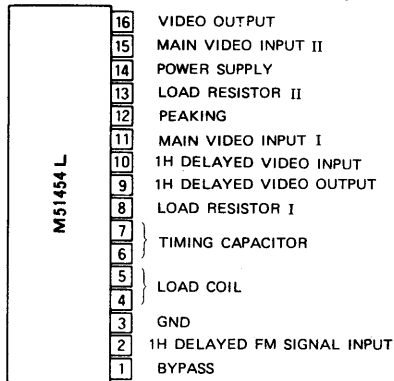
### APPLICATION

VTR video signal noise cancellation

### RECOMMENDED OPERATING CONDITIONS

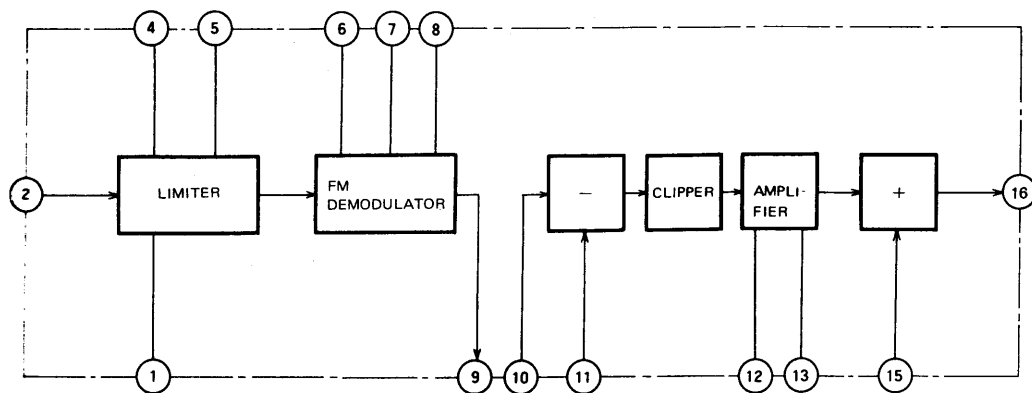
- Supply voltage range . . . . . 8~10V
- Load supply voltage . . . . . 9V

### PIN CONFIGURATION (TOP VIEW)



16-pin molded plastic ZIL

### BLOCK DIAGRAM





VIDEO SIGNAL NOISE CANCEL

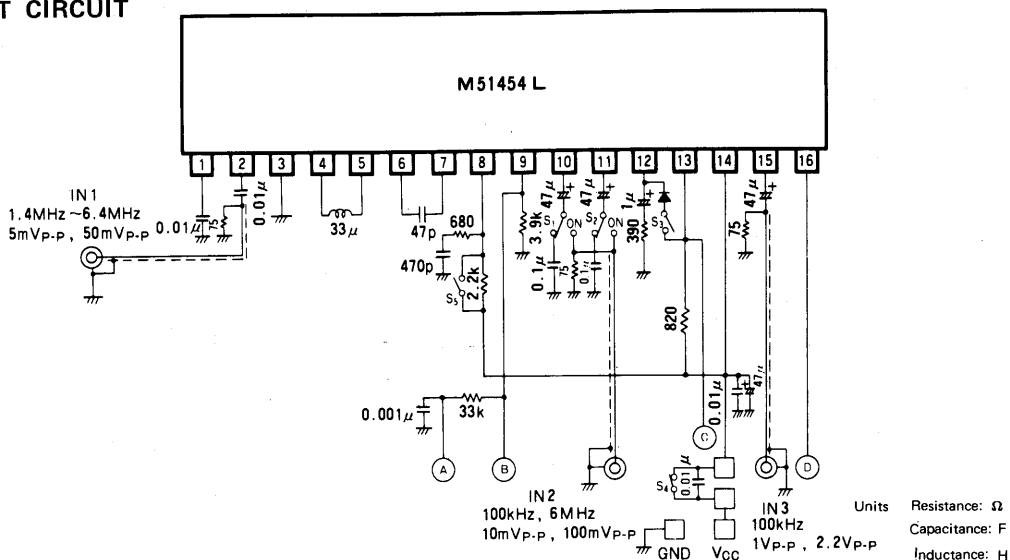
ABSOLUTE MAXIMUM RATINGS (T<sub>a</sub>=25°C, unless otherwise noted)

Symbol	Parameter	Conditions	Limits	Unit
V <sub>CC</sub>	Supply voltage		14	V
P <sub>d</sub>	Power dissipation		800	mW
T <sub>opg</sub>	Ambient temperature		-20 ~ +75	°C
T <sub>stg</sub>	Storage temperature		-40 ~ +125	°C
K <sub>θ</sub>	Thermal derating (T <sub>a</sub> ≥ +25°C)		8	mW/°C

ELECTRICAL CHARACTERISTICS

記号	Parameter	Test conditions	Limits			Unit
			Min	Typ	Max	
S <sub>DEM</sub>	FM demodulation sensitivity	f = 3.4MHz, 4.4MHz	95	115	135	mV/M
L <sub>DEM-1</sub>	FM demodulation linearity 1	f = 6.4MHz	-5	0	5	%
L <sub>DEM-2</sub>	FM demodulation linearity 2	f = 1.4MHz	-5	0	5	%
CL	Carrier leakage	f = 4MHz		-34	-26	dB
S <sub>RF</sub>	RF sensitivity	f = 4MHz	-5	0	5	mV
G <sub>V-1</sub>	Gain 1	Pin ⑩ input 10mV <sub>p-p</sub> (100kHz)	12.5	15.5	18.5	dB
G <sub>V-2</sub>	Gain 2	Pin ⑪ input 10mV <sub>p-p</sub> (100kHz)	12.5	15.5	18.5	dB
R <sub>CLP</sub>	Clipping capacity	Output ratio between pin ⑩ input 50mV <sub>p-p</sub> and 100mV <sub>p-p</sub> (100kHz)	75	85	95	%
G <sub>OS</sub>	Gain offset	100mV <sub>p-p</sub> (100kHz) input simultaneously at pins ⑩ and ⑪			20	mV <sub>p-p</sub>
G <sub>F-1</sub>	Frequency response 1	Output ratio between pin ⑩ input 10mV <sub>p-p</sub> , 100kHz and 6MHz	-3	0	0	dB
G <sub>F-2</sub>	Frequency response 2	Output ratio between pin ⑪ input 10mV <sub>p-p</sub> , 100kHz and 6MHz	-3	0	0	dB
S <sub>Q</sub>	Squelch function	Pin ⑩ input 100mV <sub>p-p</sub> (100kHz)			10	mV <sub>p-p</sub>
G <sub>V-3</sub>	Gain 3	Pin ⑮ input 1V <sub>p-p</sub> (100kHz)	-1	0	1	dB
DR	Dynamic range	Pin ⑮ input 2.2V <sub>p-p</sub> (100kHz)	-5	0	5	%
I <sub>CC</sub>	Supply current	Including current (1.9mA) of pin ⑨ external resistor (3.9kΩ)	12.5	16.7	20.1	mA

TEST CIRCUIT



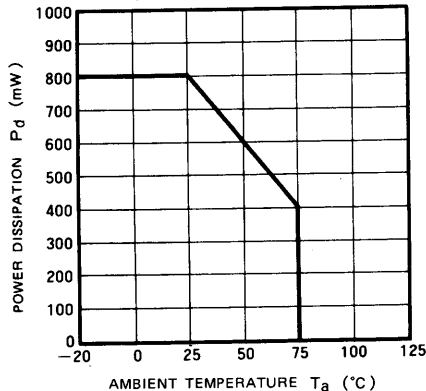
VIDEO SIGNAL NOISE CANCELER

TEST METHODS (Ta=25°C, VCC=9V, unless otherwise noted)

Symbol	S1	S2	S3	S4	S5	Method
S <sub>DEM</sub>	OFF	OFF	ON	ON	OFF	If pin ② 3.4MHz, 4.4MHz, 50mVp-p input, point(A) output voltages are made V <sub>1</sub> and V <sub>2</sub> , respectively: $S_{DEM} = (V_2 - V_1) \times 1000/2$ (mV/MHz)
L <sub>DEM-1</sub>	OFF	OFF	ON	ON	OFF	If pin ② 6.4MHz, 50mVp-p input, point(A) output voltage is made V <sub>3</sub> : $L_{DEM-1} = \left\{ \frac{V_3 - V_2}{2(V_2 - V_1)} - 1 \right\} \times 100$ (%)
L <sub>DEM-2</sub>	OFF	OFF	ON	ON	OFF	If pin ② 1.4MHz, 50mVp-p input, point(A) output voltage is made V <sub>4</sub> : $L_{DEM-2} = \left\{ \frac{V_1 - V_4}{2(V_2 - V_1)} - 1 \right\} \times 100$ (%)
C <sub>L</sub>	OFF	OFF	ON	ON	OFF	Pin ② 4MHz, 50mVp-p (less than -50dB second harmonic level) input, point(B) observed on spectrum analyzer, 8MHz : 4MHz ratio read in decibels
S <sub>RF</sub>	OFF	OFF	ON	ON	OFF	If pin ② 4MHz, 10mVp-p, 100mVp-p input, point(A) output voltages are made V <sub>5</sub> and V <sub>6</sub> , respectively: $S_{RF} = (V_5 - V_6) \times 1000$ (mV)
G <sub>V-1</sub>	ON	OFF	OFF	ON	ON	If pin ⑩ 100kHz, 10mVp-p input, point(D) signal voltage is made V <sub>1</sub> : $G_{V-1} = 20 \log_{10} \frac{V_1}{10}$ (dB)
G <sub>V-2</sub>	OFF	ON	OFF	ON	ON	If pin ⑩ 100kHz, 10mVp-p input, point(D) signal voltage is made V <sub>2</sub> : $G_{V-2} = 20 \log_{10} \frac{V_2}{10}$ (dB)
R <sub>CLP</sub>	ON	OFF	OFF	ON	ON	If pin ⑩ 100kHz, 50mVp-p, 100mVp-p input, point(D) signal voltages are made V <sub>3</sub> and V <sub>4</sub> , respectively: $R_{CLP} = \frac{V_3}{V_4} \times 100$ (%)
P <sub>in</sub>	ON	ON	OFF	ON	ON	Pin ⑩, pin ⑪ 100kHz, 100mVp-p input, point(D) output signal voltage
G <sub>F-1</sub>	ON	OFF	OFF	ON	ON	If pin ⑩ 100kHz, 6MHz, 10mVp-p input, point(D) output voltages are made V <sub>5</sub> and V <sub>6</sub> , respectively: $G_{F-1} = 20 \log_{10} \frac{V_6}{V_5}$ (dB)
G <sub>F-2</sub>	OFF	ON	OFF	ON	ON	If pin ⑩ 100kHz, 6MHz, 10mVp-p input, point(D) output voltages are made V <sub>7</sub> and V <sub>8</sub> , respectively: $G_{F-2} = 20 \log_{10} \frac{V_8}{V_7}$ (dB)
P <sub>out</sub>	ON	OFF	ON	ON	ON	Pin ⑩ 100kHz, 100mVp-p input, point(D) output voltage
G <sub>V-3</sub>	OFF	OFF	ON	ON	ON	If pin ⑮ 100kHz, 1Vp-p input, point(D) output voltage is made V <sub>1</sub> : $G_{V-3} = 20 \log_{10} V_1$
DR	OFF	OFF	ON	ON	ON	If pin ⑮ 100kHz, 2.2Vp-p input, point(D) output voltage is made V <sub>2</sub> : $DR = \left( 1 - \frac{V_2}{2.2V_1} \right) \times 100$ (%)
	OFF	OFF	OFF	OFF	OFF	

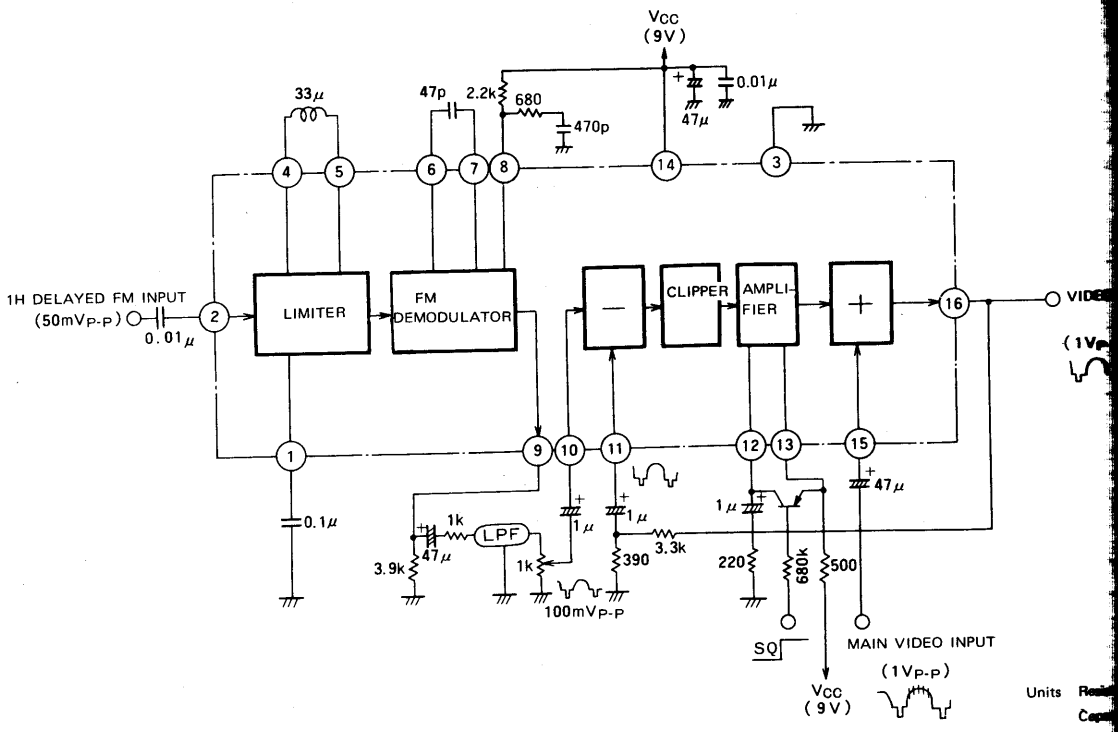
TYPICAL CHARACTERISTICS (Ta=25°C, VCC=14V, unless otherwise noted)

THERMAL DERATING  
(MAXIMUM RATING)



**VIDEO SIGNAL NOISE CANCELLATION**

**APPLICATION EXAMPLE**



Units Res  
 Cap